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Corner Defected Multi-Resonance Ultra-Wideband Rectangular Patch Antenna

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Abstract

Objectives: The main objective of this article is to present a novel corner defected ultra wideband (UWB-3.1 to 10.6 GHz) rectangular patch with a partial ground plane to enhance the bandwidth with multi-resonance characteristics. By defecting the corners, the original rectangle patch will be converted into defected shape of proposed one to satisfy the objective. **Methods:** A FR-4 substrate with height (h) of 1.6 mm, dielectric constant (ϵ_r) of 4.3 and the loss tangent of 0.025 was used to develop the required antenna. The antenna was excited by a 50 Ohm microstrip line. This antenna was designed, simulated and analyzed by using MS-CST (Microwave Studio – Computer Simulation Technology) to attain the outcomes such as return loss ($S_{11} < -10$ dB), voltage standing wave ratio (VSWR < 2), improved gain, Omni-directional radiation patterns, and enhanced bandwidth. **Findings:** It exhibits a fractional bandwidth (FBW) of 152.9 % between 2 GHz and 15 GHz. The antenna resonates at multiple frequencies owing to truncation of corners and produces resonant frequencies of 3.87 GHz and 11.36 GHz with the corresponding return losses of -26.03 dB and -34.87 dB respectively in the simulation point of view. The antenna was fabricated and tested. The measured and simulated results of the desired antenna were compared, and they show good agreement. **Applications:** The proposed antenna is suitable for UWB, WiMAX (Worldwide Interoperability for Microwave Access), C and X – band communications.

Keywords: Ultrawideband; Rectangular patch; Bandwidth; Return loss; VSWR; Gain

1 Introduction

The Federal Communications Commission (FCC) had released the unlicensed frequency band, 3.1-10.6 GHz (7.5 GHz spectrum) for ultra-wideband (UWB) communication systems in Feb.14, 2002⁽¹⁾. One of the main challenges in designing of a compact patch antenna in UWB communication systems is to miniaturize the antennas

with wide impedance bandwidth while still maintaining high radiation efficiency throughout the complete operating frequency band. Microstrip patch antennas are the most useful structures due to low cost, low profile, ease of fabrication & implementation, and conformability to host surfaces i.e. planar and non-planar. One of the key drawbacks of patch antennas is their narrow bandwidth⁽²⁾. Therefore, for some of the wireless applications such as satellite communications; it is desirable to enhance the impedance bandwidth without affecting the performance of patch is a crucial task for researchers. Different bandwidth enhancement techniques have been presented in the literature; some of them are applied to radiating element or ground plane or some to both. A rectangular UWB patch with defected ground structure was presented in⁽³⁾. In this, the upper corners of the patch are square notched and lower corners are rectangular notched to enhance the bandwidth. This antenna produced a bandwidth of 3.5-14.9 GHz with percentage bandwidth of 123.91%. A UWB rectangular patch with defected ground structure (DGS) was addressed in⁽⁴⁾. The bandwidth of this antenna could be increased by one round cut at each corner of the patch with one slot in the ground plane. This antenna provided a BW of 3.42-11.7 GHz. In⁽⁵⁾, a rectangular UWB antenna, which is capable of supporting wide BW of 3-10.26 GHz by adding suitable slits and slots in the patch and ground, was proposed. An antenna for wireless applications was introduced in⁽⁶⁾ which was also acted as UWB antenna by providing a polygonal slot in the plane of ground. The enhancement of BW can be achieved in⁽⁷⁾ by utilizing CMA (Characteristic Mode Analysis) technique along with the structure of defected ground. The bandwidth of an antenna can also be increased by using a circular patch with a defective structure of ground⁽⁸⁾. Multiband operation can be attained by the octagonal patch with a slot⁽⁹⁾. UWB can also be implemented by semi-circular patch with fractal and defected ground⁽¹⁰⁾. A pentagonal patch with DGP (defected ground plane) was invented for ultra wideband⁽¹¹⁾.

A UWB patch antenna for C, X- band and for improving the BW was addressed in⁽¹²⁾. A rectangular patch was built for WLAN and WiMAX in⁽¹³⁾, which was also meant for enhancing the BW by introducing slot and notches in the patch. Gap coupling is the technique used for improving the impedance BW of a patch⁽¹⁴⁾. A wearable U shaped patch was designed for the applications of C, WLAN and HIPERLAN/2⁽¹⁵⁾. UWB and multi-wireless applications were attained by a circular patch with a slot⁽¹⁶⁾. An ultra-wideband slotted hexagonal antenna for BW enhancement was described in⁽¹⁷⁾. The operation of UWB was depicted by a stacked patch⁽¹⁸⁾ and decagonal patch⁽¹⁹⁾. In⁽²⁰⁾, a broadband patch was represented, and increase in bandwidth could be obtained by cutting two circular slots on the patch and truncating the corners of the square patch. An M & A - shaped radiating patches with modified ground by cutting rectangular slots^(21,22), a compact modified rectangular patch and a planar antenna with modified patch and defected ground plane^(23,24) were addressed for UWB communications. A corner truncated patch with slotted ground⁽²⁵⁾ can also enhance the impedance bandwidth.

In this paper, the proposed antenna not only satisfies both UWB and BW enhancement but also yields multiple resonant characteristics at various frequencies in the measured point of view. The idea of designing of this antenna is derived from^(3,4). In this context, the upper edge corners are square truncated and lower edge corners are arc truncated to provide the larger bandwidth with multi-resonance properties. The papers reported in the above survey of literature did not enhance the bandwidth more than 11.4 GHz while the proposed one demonstrates a FBW of 152.9% lies between 2 and 15 GHz. And it can be employed for UWB, WiMAX, C and X-band applications.

2 Design and Geometry of Antenna

The schematic representation and procedure of evaluation of corner defected patch (CDP) is depicted from Figure 1 a-f. In this work a rectangular patch having the volume of (L_p mm \times W_p mm \times t_p mm) was designed on top of the substrate with the dimensions of (L_s mm \times W_s mm \times h mm), where L_p = Length of the rectangular patch, W_p = Width of the patch, t_p = Thickness of the patch, L_s = Length of the substrate, W_s = Width of the substrate and h = Height of the substrate. The CDP was fed by a microstrip line has an area of (L_f mm \times W_f mm), where L_f = Length of the feed line and W_f = Width of the feed line. A partial ground plane, L_g mm \times W_g mm, was printed in the rare side of the substrate. Where L_g = Length of the ground plane and W_g = width of the ground plane. The top corners of (x mm \times y mm), where x = Horizontal length of top corners notch and y = Vertical length of top corners notch and $x = y = 2.5$ mm, and the bottom corners of radius 'a' mm were etched. All these parameter values are shown in below Table 1. Figure 1 (e) shows the configuration and dimensions of the desired antenna, which is fabricated on lossy FR-4 substrate with relative permittivity of $\epsilon_r = 4.3$, permeability of $\mu_r = 1$, loss tangent of $\tan \delta = 0.025$ and the thickness of the substrate of $h = 1.6$ mm. The shape and dimensions of this antenna is based on the antennas presented in^(3,4). The antenna is designed at the design frequency of $f_o = 8.5$ GHz.

The basic rectangular patch is designed by using the following formulae as

$$W_p = \frac{c}{2f_o} \sqrt{\frac{2}{\epsilon_r + 1}} \quad (1)$$

Where W_p = Width of the rectangular patch, f_o = Design frequency ϵ_r = Relative permittivity of the dielectric.

The effective di-electric constant (ϵ_e) of the dielectric (FR-4) is given by

$$\epsilon_e = \frac{\epsilon_r + 1}{2} + \frac{\epsilon_r - 1}{2} \left[\frac{1}{\sqrt{1 + 12 \frac{h}{W_p}}} \right] \quad (2)$$

Where h = Height of the substrate, and.

$$(1 < \epsilon_e < \epsilon_r)$$

The actual length (L_p) of the rectangular patch can be determined by

$$L_p = \frac{c}{2f_o \sqrt{\epsilon_e}} - 2\Delta L_p \quad (3)$$

$$(\Delta L_p) = 0.412h \frac{(\epsilon_e + 0.3) \left(\frac{W_p}{h} + 0.264 \right)}{(\epsilon_e - 0.258) \left(\frac{W_p}{h} + 0.8 \right)} \quad (4)$$

Where ΔL_p = Extended incremental length of the patch. ΔL_p can be calculated by

The effective length (L_e) of the patch can be computed by

$$L_e = L_p + 2\Delta L_p \quad (5)$$

For the dominant mode (TM_{010}), the design frequency of rectangular patch is expressed as

$$(f_o)_{010} = \frac{c}{2(L_p + 2\Delta L_p) \sqrt{\epsilon_e}} \quad (6)$$

Table 1. Parametric Dimensions of Antenna

Parameter	Value(mm)
L_s	35
W_s	30
h	1.6
L_p	14.5
W_p	15
t_p	0.1
L_f	13.5
W_f	3.2
L_g	12.5
W_g	30
x	2.5
y	2.5
a	2.8

3 Results and Discussion

3.1 Discussion of simulation results

The plain structure of antenna shown in Figure 1 a resonates at the frequency 3.95 GHz and the corresponding return loss (RL) and VSWR are -18.6 dB and 1.2. This structure operates in the frequency range of 3.1-11.1 GHz. The top left corner defected

structure is shown in Figure 1b, for which $f_0 = 4.1$ GHz, and -18.5 dB and 1.2 are the RL and VSWR at design frequency. It provides the BW of 7.9 GHz (3.2-11.1 GHz). In addition to top left corner, the top right corner defected antenna is shown in Figure 1c, which produces the resonant frequency of 4.2 GHz and its S_{11} and VSWR are -19.3 dB and 1.2. This design almost produced the same BW as the previous structure. Besides the top corners, the lower left corner is arc defected is shown in Figure 1d. The resonant frequency, S_{11} , VSWR and bandwidth of this antenna are 6.9 GHz, -61.5 dB, 1.001 and 8.9 GHz (3.2-12.1 GHz). The desired antenna is shown Figure 1e. This antenna has been presented $S_{11} = -33.9$ dB, VSWR = 1.04 at the resonant frequency of 11.3 GHz. The antenna is operating in the frequency band of 3.1 GHz to 14.6 GHz (BW = 11.5 GHz) which includes UWB, C band (4-8 GHz) and X-band (8-12.4 GHz). This antenna provides a larger BW and is operating at dual bands i.e. at 3.87 GHz (Wi-MAX) and 11.36 GHz (Satellite down link) frequencies. A partial ground plane is shown in Figure 1f is used for all the designs. These results are shown in Figure 2 a & b. All the simulated results of evaluation structures of patch are tabulated in Table 2. The simulation and measured results of proposed antenna are shown in Table 3. From this comparison, it is concluded that the desired antenna produces a peak gain of 4.69 dB, larger BW in addition to UWB characteristics.

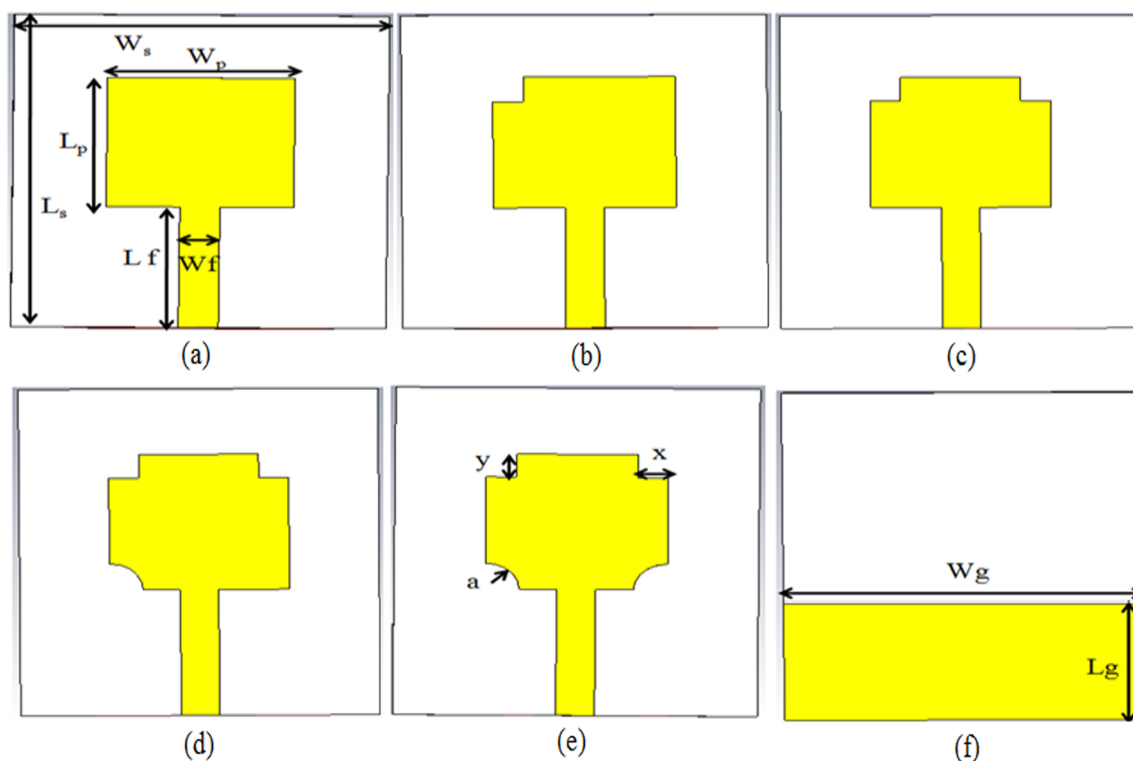


Fig 1. (a) Design-1 (b) Design-2 (c) Design-3 (d) Design-4 (e) Design-5 (Proposed) (f). Partial ground plane (Bottom view).

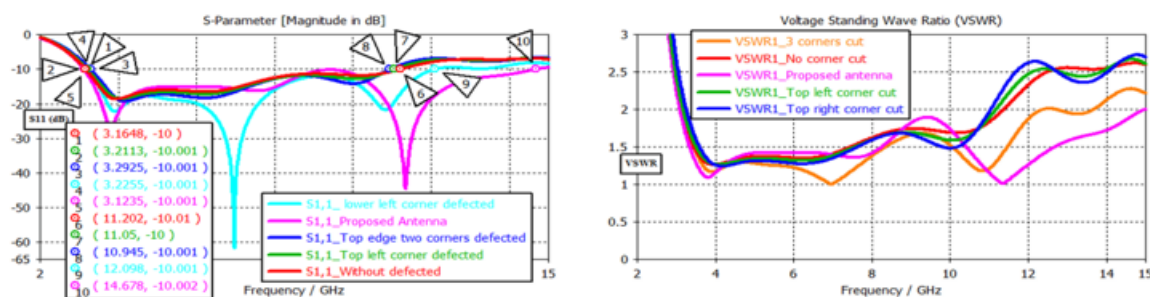


Fig 2. (a) Return loss plots for all designs (b). VSWR plots for all designs

Table 2. Comparison of simulation results for evaluation structures

Antenna	Resonant Freq.(GHz)	S11(dB)	VSWR	BW(GHz)	Gain at 8.5 GHz
Design 1	3.9	-18.6	1.2	3.1-11.1	4.48
Design 2	4.1	-18.5	1.2	3.2-11.1	4.86
Design 3	4.2	-19.3	1.2	3.2-11	4.66
Design 4	6.9	-61.5	1.001	3.2-12.1	4.84
Design 5	3.87 11.36	-26.03 -34.87	1.04	3.1-14.6	4.69

Table 3. Simulated & Measured results of proposed antenna

Parameter	Simulated	Measured
S11(dB)	-34.87	-37
VSWR	1.04	1.1
Operating range (GHz)	3.1-14.6	2-15
BW(GHz)	11.5	13
Peak Gain(dB)	4.69	3.98
Resonant Frequency(GHz)	11.3	7.8
Radiation patterns	Omni-directional	Omni-directional

3.2 Radiation patterns & Surface current distribution

The three-dimensional gain patterns and polar plots of E & H plane patterns at 6, 9 and 10.5 GHz frequencies are shown from Figure 3 a-f, from these patterns the proposed antenna radiates nearly Omni-directional patterns. The distribution of surface current at 6, 9 and 10.5 GHz are shown in Figure 4 a–c. The electromagnetic signals are radiating from edges of the patch and ground plane. The maximum values of surface current are 50.1, 46.2 and 62.1 A/m for 6, 9 and 10.5 GHz frequencies respectively. It is evident from Figure. 4 that strong current distribution is observed along the edges of the patch and the ground plane and the current flow is dominant on the feed line also. The maximum current observed at the higher frequencies only but not at lower frequencies.

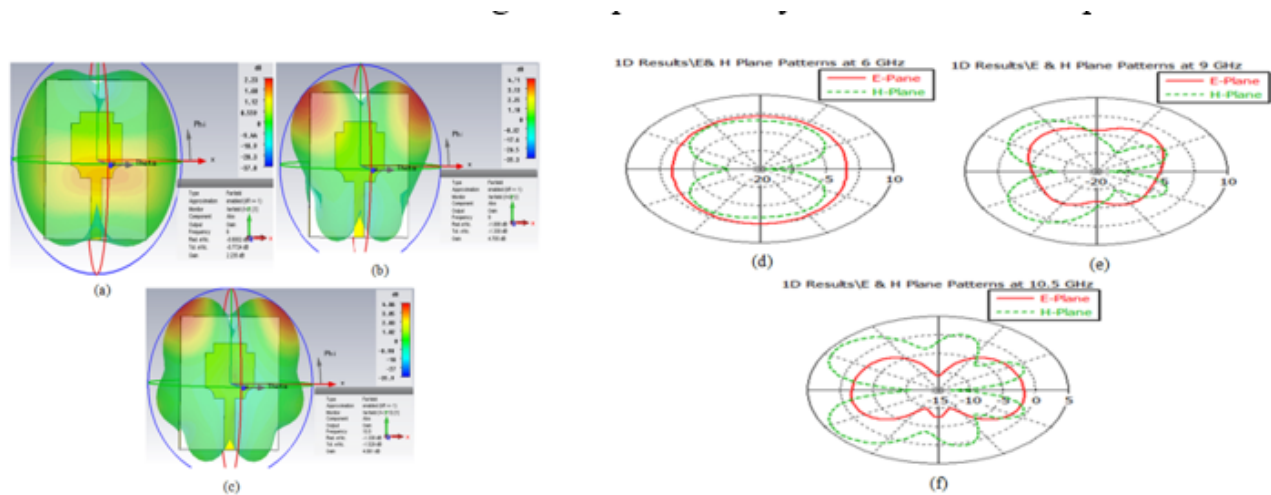


Fig 3. Three-dimensional gain pattern at (a) 6 GHz, (b) 9 GHz (c) 10.5 GHz frequencies. Polar E-plane and H-plane patterns at (d) 6, (e) 9, (f) 10.5 GHz frequencies respectively

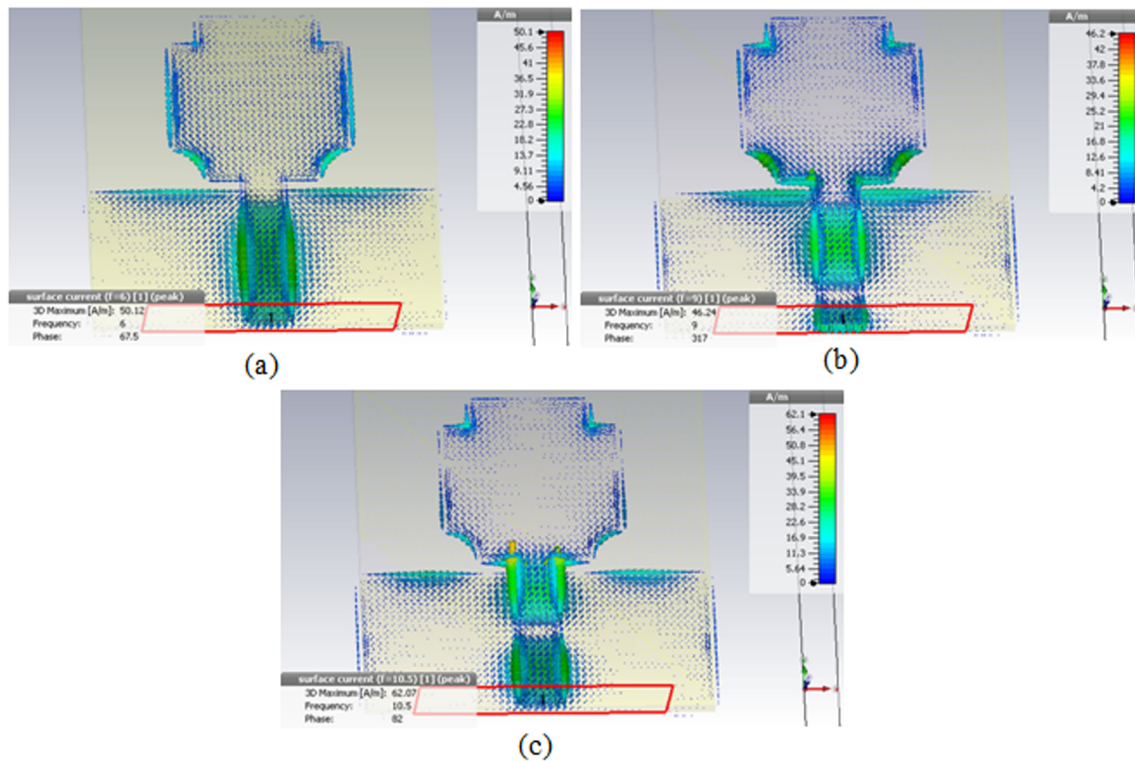


Fig 4. Simulated surface current distribution at (a) 6 GHz,(b) 9 GHz & (c) 10.5 GHz frequencies.

3.3 Experimental results and discussion

A vector network analyzer is utilized to measure the performance of the desired antenna such as return loss, VSWR, and impedance BW. An anechoic chamber is used to measure the radiation patterns. The prototype of the fabricated antenna was under testing is represented in Figure 5. The measured return loss and VSWR of the desired antenna are shown in Figure 6 a & b. It was observed from Figure 6 that the measured return loss is less than -10 dB and VSWR less than 2, which show a good agreement with the simulated results as shown in Figure 2. From the measured return loss plot, it is identified that the antenna resonates at multiple frequencies. The fabricated antenna provides a measured BW of 13 GHz (2-15 GHz), which covers a whole ultra-wideband (3.1-10.6 GHz), Wi-MAX (3.87 GHz), C band (4-8 GHz), X-band (8-12 GHz) and lower part of the K_u band (12-15 GHz). Figure 7a–c show the measured polar far-field gain patterns in vertical (E) and horizontal (H) plane patterns at 6, 9 and 10.5 GHz frequencies respectively. It is observed that the antenna produced a nearly good Omni-directional patterns at all frequencies in the vertical and horizontal planes. These patterns, as expected, are suitable for UWB, WiMAX, C, X and K_u band applications.



Fig 5. Fabricated antenna prototype in testing.

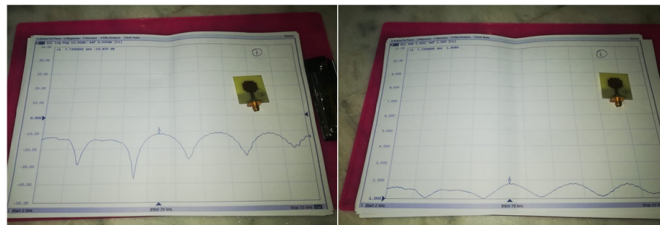


Fig 6. (a). Measured Return loss (dB); (b). Measured Voltage standing wave ratio (V.S.W.R)

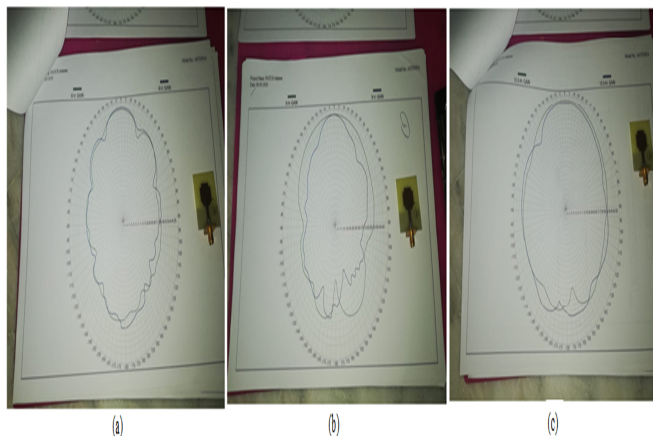


Fig 7. Measured gain patterns at(a) 6 (b) 9 (c) 10.5 GHz frequencies

Gain is an essential parameter in the design of ultra-wideband antenna. The simulated peak gain is 4.69 dB and measured peak gain is 3.98 dB. The proposed antenna was compared to some of the references in literature as shown in Table 4. The desired antenna has been produced an absolute BW of 13 GHz and fractional BW of 152.9%, which is the highest of all existing ones in the literature. By means of arc truncation in the lower edge corners of the patch, the bandwidth is drastically increased. Figure 8 shows that the comparison of simulated and measured gain plot, which shows that gain, varies from 1.8 to 3.9 dB and it maintains nearly constant gain but less than 5 dB, which confirms that the measured patterns are Omni-directional. The maximum radiation of this antenna is 92.9% at 2 GHz and varies between 64.3% to its maximum value between the frequencies 2 to 15 GHz that is shown in Figure 9. Therefore, the proposed antenna is the suitable one for UWB applications.

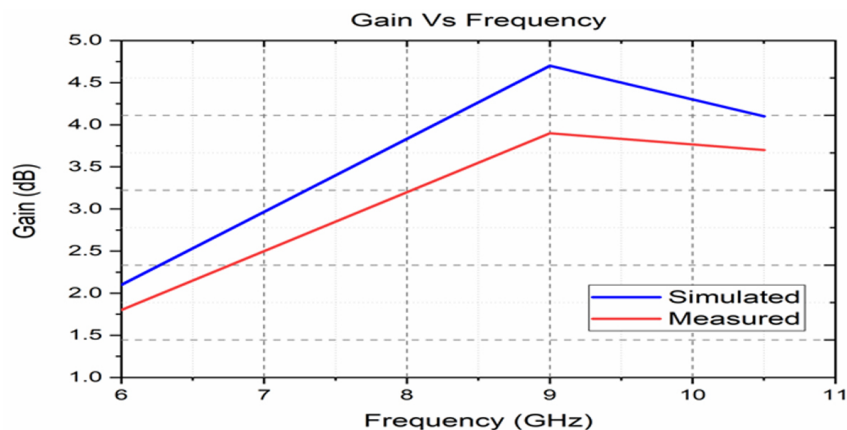


Fig 8. Comparison plot of simulated & measured gain values

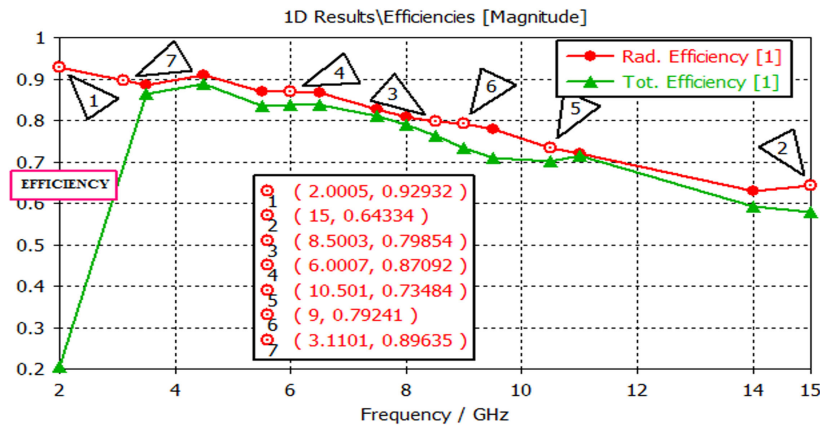


Fig 9. Radiation efficiency of proposed antenna (Simulation results)

Table 4. Comparison of Proposed antenna with some of the existing Antennas in the References.

Ref.	Antenna Size(W×L×h),mm ³	Substrate +Ground+ Feed types	-10dB Operating Range(-10 dB B.W)GHz	F.B.W (%)	Gain(dB)
3	36×34×1.6	FR-4+D.G.S ¹ +M.F.L ²	3.5-14.9 (11.4)	123.91	5
4	30×35×1.6	FR-4+D.G.S ¹ +M.F.L	3.42-11.7(8.28)	109.5	6
5	36×34×1.6	FR-4+P.G.P+M.F.L	3-10.26 (7.26)	109.5	5
6	45×45×1.565	FR-4+D.G.S+M.F.L	2.08-6.64(4.56)	104	5.4
7	50×25×0.17	Foam+D.G.S+Coaxial feed	1.82-3.02(1.2)	49.09	3
8	38×48×1.6	FR-4+P.G.P+M.F.L	2.5-10.6 (8.1)	123.6	8.4
10	47×30×1.64	FR-4+D.G.S+M.F.L	4.395-10.184 (5.789)	79.4	3.84
11	17.59×30×1.6	FR-4+D.G.S+M.F.L	2.66-10.82 (8.16)	121	3.38
12	24×24×1.6	FR-4+P.G.P+M.F.L	3.53-11.64(8.11)	106.92	7.2
13	37×30.8×1.6	FR-4+P.G.P+M.F.L	2.08-3.99(1.91)	62.9	4.7
14	40×40×1.6	Glass Epoxy+P.G.P+M.F.L	2.093-6.105(4.012)	97.88	6.725
15	42×43.5×1	Jeans Textile+D.G.S+M.F.L	4.8-9(4.2)	60.86	5.19
17	29×28×1.6	FR-4+D.G.S+M.F.L	5.73-10.8(5.07)	61.34	6
19	35×35×1.6	FR-4+D.G.S+M.F.L	2.3-12.8(10.5)	139	5
22	25×25×1.6	FR-4+D.G.S+M.F.L	2.9-11.5(8.6)	119.4	8
Proposed Antenna	30×35×1.6	FR-4+P.G.P+M.F.L	2-15 (13)	152.9	4.69(simulated) 3.98(measured)

D.G.S¹=Defected Ground structureM.F.L²=Microstrip Feed LineP.G.P³=Partial Ground Plane

4 Conclusion

A corner defected multi-resonant UWB patch with a partial ground plane was designed, simulated and analyzed by CST and compared with the experimental results. This antenna yields a wide spectrum of 13 GHz i.e. 2-15 GHz which includes UWB, WiMAX, C and X-band (8-12.4 GHz). This antenna has given a return loss of < -10 dB, VSWR of less than 2 throughout the operating band. The antenna almost maintains a consistent gain within the operating frequency and is less than 5 dB. This antenna radiates multi resonance and Omni-directional radiation characteristics. Therefore, the proposed antenna is suitable for UWB, WiMAX, C, X-band and lower portion of K_u band applications.

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